

AUSTRALIA

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PROVISIONAL SPECIFICATION

Invention Title:
"PHASE SHIFTER"

This invention is described in the following statement:

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Field of Invention

This invention relates to antennas and in particular to an arrangement to electrically down-tilt the electromagnetic wave pattern associated with a transmit antenna array, or electrically re-orient a receive antenna array.

5 Background of Invention

It is sometimes desirable to adjust the orientation of the electromagnetic wave pattern of a transmit antenna array, particularly a downward adjustment, typically 0 degrees to 15 degrees below horizontal, when the antenna is located at a higher altitude than other antennas that communicate with the transmit antenna array. The downward adjustment of the radiation pattern alters the coverage area and may enhance communication with mobile users situated in shadowed areas below the transmit antenna array.

Besides actually mechanically tilting the entire antenna assembly, it is known to electrically down-tilt the radiation pattern by controllably varying the relative phase or phases between two or more radiating elements of the antenna array.

One known method by which the relative phase between two or more radiating elements can be changed is to change the relative lengths of respective transmission lines connecting the antenna's common feed point to each element of the antenna array. Typically, various predetermined lengths of jumper cable are provided which are selectively connected between the common feed and each element to obtain a desired down-tilt. The jumper cables include co-axial connectors to facilitate connection. Furthermore, if stripline is used to connect the common feed point to the respective elements of the antenna array, some form of transition means is required to couple the jumper cable's co-axial connections to the stripline. A disadvantage of this known method is that it is expensive, unreliable and susceptible to the generation of intermodulation products.

Another known method by which the relative phase between two or more radiating elements can be changed is to change the propagation velocity of the transmission line connecting the common feed point to at least some of the elements of the antenna array. Typically, this latter method is achieved by selectively changing the dielectric constant of the transmission line dielectric. If the transmission line is in the form of a conductive strip, the propagation velocity

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thereof is changed by introducing a dielectric material between the strip and its associated ground plane.

It is, however, well understood that the introduction of dielectric material under such a conductive strip causes the strip's normal impedance to be
5 disturbed. For example, if a conductive strip having a certain width is spaced above a ground-plane at a certain distance such as to present a 50 ohm impedance, the introduction of a dielectric material between the conductive strip and the ground-plane will reduce the value of this impedance to a value that depends upon the effective dielectric constant of the dielectric material. The
10 resulting impedance mismatch would cause a degradation of return-loss performance of the antenna.

Australian Patent No. 664625 discloses an arrangement of an adjustable phase shifter comprising dielectric phase shifter elements moveably interposed between conductive strips that couple radiating elements, and a common ground-
15 plane. The phase shifter elements are of a characteristic configuration which avoids disturbing the normal impedance during adjustment. This known arrangement, however, requires that respective phase shifter elements be located between each active strip and the conductive ground-plane. Such an arrangement imposes constructional disadvantages as well as limitations to the
20 range of phase shift produced, which consequently imposes limits to the range of tilt.

Australian Patent Application No. 14278/99 discloses an arrangement of an adjustable phase shifter comprising a transmission line in the form of a printed circuit board supporting conductive tracks on one side thereof, and a ground-
25 plane spaced below the other side thereof. A moveable dielectric element is arranged adjacent the conductive tracks. The moveable dielectric element is provided with a plurality of teeth along opposite edges for selectively overlapping the conductive tracks. This arrangement provides an adjustable phase shifter having stable impedance characteristics and a relatively large phase shift as
30 compared with the prior art.

New radio frequency bands have been allocated to provide more channels for the rapidly increasing cellular mobile telephone usage. Instead of having separate base station antennas for different bands, it is desirable to provide

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multiple-band antennas. For example, it may be desirable to combine the 1710 – 1880 MHz DCS band with the 1920 – 2170 MHz UMTS band, with an overall bandwidth of 460 MHz, which is wider than previous systems. In order to electrically down-tilt the radiation pattern of such a multiple band antenna, a phase shifter is required that has a wider frequency range than the aforementioned prior art phase shifter arrangements can accomplish.

Summary of Invention

It is an object of the present invention to provide an adjustable radio frequency phase shifter arrangement having a wider operating frequency range than prior art arrangements.

According to the invention there is provided a phase shifter element arranged to selectively vary the effective dielectric constant of a section of transmission line thereby changing the propagation velocity of said transmission line and varying the phase of signals of desired frequencies or frequency range passing through said transmission line, said phase shifter element comprising a movable planar dielectric member of predetermined dielectric constant adjacent said transmission line, said planar dielectric member being provided with three or more discrete interactive dielectric segments extending from at least one edge thereof to moveably overlap the adjacent transmission line, wherein optimum dimensions of each said interactive segment and optimum widths of gaps defined by opposite edges of adjacent segments are determined by computer optimisation means, such that the aggregate reflection of said signals passing through said transmission line is minimised.

The present invention is based on the concept that, as phase shift is generally proportional to the length of added or inserted dielectric, if the total dielectric length required for one particular phase shift is broken up into a plurality of segments, then it is possible to optimise all of the lengths and spacings of those segments in such a way as to reduce the aggregate reflection at several frequencies, or over a range of frequencies. Thus for a specified frequency range and for one required electrical phase shift, computer optimisation is used to determine the optimum lengths of a plurality of dielectric segments and the spacings between them. The same procedure is repeated for other phase shift values to produce an optimum profile for the movable dielectric element.

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In order that the invention may be readily carried into effect, embodiments thereof will now be described in relation to figures of the accompanying drawings, in which:

Brief Description of the Drawings

5 Figure 1 is a plan view of a planar dielectric element provided with a plurality of characteristic spaced segments extending therefrom.

Figure 2 is a side view of Figure 1.

Figure 3 is a plan view of a phase shifter arrangement incorporating the dielectric element of the present invention.

10 Figure 4 is a side view of the arrangement shown in Figure 3.

Figure 5 is a schematic layout of an antenna array incorporating the phase shifter arrangement shown in Figure 3.

Figure 6 is a plan view a further embodiment of a phase shifter arrangement incorporating the dielectric element of the present invention.

15 Figure 7 is a schematic layout of an antenna array incorporating the phase shifter arrangement shown in Figure 6.

Referring to Figures 1 and 2 of the drawings there is shown a planar dielectric element 1 comprising a rectangular body section 2 and five segments, 3,4,5,6 and 7 extending from a major edge of body section 2. The segments are separated by four air gaps 8,9,10 and 11. The segments lie in the same plane as the body section. To improve structural rigidity of the dielectric element, the air gaps may be replaced by a dielectric material of a different dielectric constant to that of the material of the dielectric element 1. Alternately, the air gaps may be replaced by thinner portions of the same material as the dielectric element.

20 As shown in Figures 3 and 4, dielectric element 1 is slidably mounted and adjacent to the top surface of a PCB distribution element comprising a planar dielectric circuit board 12 supporting a conductive track 13 on a first surface 12a thereof. The conductive track and the dielectric circuit board form a transmission line whose distal ends terminate at respective terminals T and B. The distribution element is supported in a spaced relationship with a conductive ground plane 14. The dielectric circuit board's second surface 12b and the ground plane face one another. Alternately, the second surface of the circuit board and the ground plane can be contiguous (not shown). The movable dielectric element 1 is supported

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above the first surface 12a of circuit board 12 in a linearly slidable manner by two parallel rods 15,16 attached to the ground plane. It will be understood that the movable dielectric element will have the effect of varying the phase whether it is adjacent the first surface 12a or the second surface 12b, although the phase shift achieved by each arrangement will be different; the movable dielectric element will have a greater effect when adjacent the second surface 12b, i.e., interposed between surface 12b and the ground plane 14.

The greater the number of segments extending from the body section of the element 1, the lesser the aggregate signal reflection caused by the dielectric element. However, the greater the number of segments requires a greater overall length of the element. But the length of the element has to be taken into account when operatively associating the element with an antenna array, to avoid constructional problems. Therefore, the choice of the number of segments is a compromise between electrical performance and practical dimensions.

If, as shown in Figure 1, the element 1 has five segments (and correspondingly four gaps) then an equivalent electrical circuit comprises nine equivalent transmission line sections in series. For one specific phase shift and desired frequency range, the lengths of those nine elements are adjusted (optimised) by using known commercially available radio frequency circuit analysis and optimisation software, to simultaneously achieve the desired phase shift and minimise the aggregate signal reflection due to the presence of the dielectric segments. The lengths of the equivalent transmission lines then represent the optimum lengths of the dielectric segments, and the optimum width of the gaps, adjacent to or overlapping the physical transmission line conductor for that particular phase shift. The same process is then repeated again and again for different phase shift values desired. Then, to construct the means by which a linear transverse movement of the dielectric element is converted to a variable phase shift, the various lengths of dielectric segments and widths of gaps overlapping the transmission line are joined to produce the required profiles of the segments. From these profiles, information is derived in a known manner and used in suitable numerically controlled cutting equipment to produce the complex shaped segments of the dielectric element.

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element 17 is slidably mounted and adjacent to one surface of a planar PCB distribution element 18 that supports two conductive tracks 19 and 20. The conductive tracks and the dielectric circuit board form a transmission line network for splitting a radio frequency signal applied to a signal input terminal I into three paths that terminate respectively in three terminals T, B and C for feeding the input signal to the top (T), Bottom (B) and Centre (C) sections of a three section antenna array (Fig. 7). The distribution element 18 is supported in a spaced relationship with a conductive ground plane 20; the planar dielectric circuit board's other surface and the ground plane facing one another. The movable dielectric element 17 is supported in a linearly slidable manner by two parallel rods 21 and 22 attached to the ground plane 20.

20 arrangements could be adapted, such as, for example, rotational arrangements. Further, remotely controlled servomechanisms could be adapted to move the dielectric element.